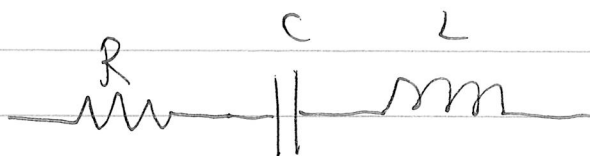


Q (a)



$$Z = R + \frac{1}{i\omega C} + i\omega L$$

series combination

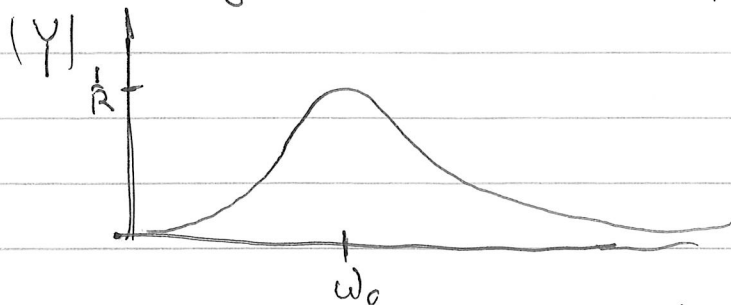
$$Z = R + i\left(\omega L - \frac{1}{\omega C}\right)$$

$$|Z| = \sqrt{R^2 + \left(\omega L - \frac{1}{\omega C}\right)^2}$$

$$|Y(\omega)| = \frac{1}{|Z|} = \frac{1}{\sqrt{R^2 + \omega^2 \left(1 - \frac{\omega_0^2}{\omega^2}\right)^2}} \quad \& \quad \omega_0^2 = \frac{1}{LC}$$

$$Y(\omega) = \frac{1}{\sqrt{R^2 + (\omega L)^2 \left[1 - \frac{\omega_0^2}{\omega^2}\right]}}$$

Y max at $\omega = \omega_0 (= 1/\sqrt{LC})$
where only R matters $Y_{\max} = \frac{1}{R}$



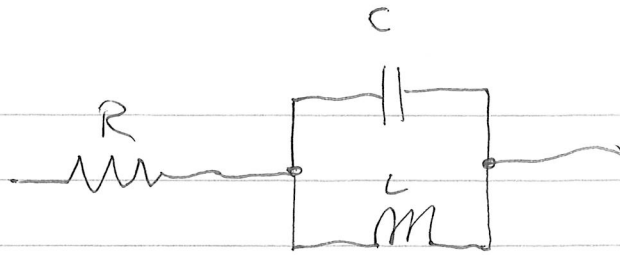
At low freq's capacitor $|Z_C| = \frac{1}{\omega C}$ becomes large, making Y small

At high freq's inductor $|Z_L| = \omega L$ becomes large, making Y small

$$Y_{\max} = \frac{1}{R} \quad \text{max. current} = I = Y_{\max} V_{i\omega}$$

so $I_{\max} = \frac{V_0}{R} e^{i\omega t}$

① (B)



Treat C & L in parallel first

$$Z_{LC} = Z_{par} = \left(\frac{1}{Z_C} + \frac{1}{Z_L} \right)^{-1}$$

$$= \left(i\omega C + \frac{1}{i\omega L} \right)^{-1}$$

$$= (i\omega C)^{-1} \left[1 + \frac{1}{i^2 \omega^2 LC} \right]^{-1}$$

$$Z_{LC} = \frac{-i}{\omega C} \left[1 - \frac{\omega_0^2}{\omega^2} \right]^{-1}$$

$$\omega_0^2 = \frac{1}{LC} \text{ again}$$

$$= \frac{-i}{\omega C} \frac{1}{1 - \omega_0^2/\omega^2}$$

$$\text{Total } Z = R + Z_{LC}$$

$$= R - i \left(\frac{1}{\omega C} \frac{1}{1 - \omega_0^2/\omega^2} \right)$$

$$|Z| = R^2 + \frac{1}{\omega^2 C^2} \frac{1}{(1 - \omega_0^2/\omega^2)^2}$$

As $\omega \rightarrow \omega_0$ the 2nd term $\rightarrow \infty$ $Z \rightarrow \infty$

As $\omega \rightarrow 0$ L is a short circuit $Z \rightarrow R$

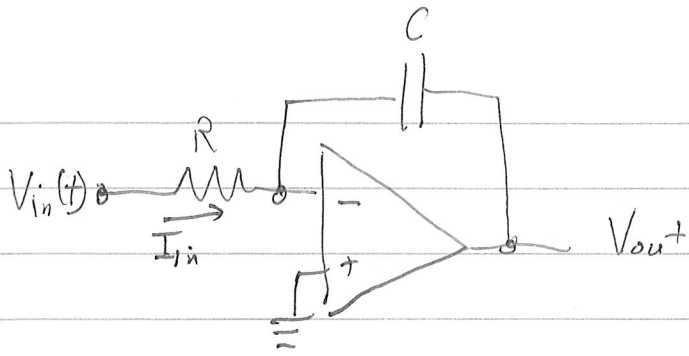
As $\omega \rightarrow \infty$ C is a short circuit $Z \rightarrow R$



$$I_{max} = \frac{V_0 e^{i\omega t}}{R} \text{ at very low or high } \left(\begin{matrix} \omega \ll \omega_0 \\ \omega \gg \omega_0 \end{matrix} \right) \text{ freq}$$

$$I_{min} = 0 \text{ at } \omega = \omega_0 = \frac{1}{\sqrt{LC}}$$

②



G.R.
 $V_+ = V_-$

Since the non-inverting pin (+) is connected to ground, the op-amp will adjust V_{out} , trying to make the - pin also ground.

$$\text{So } I_{in}(t) = \frac{V_{in}(t) - 0}{R} = \frac{V_{in}(t)}{R}$$

G.R.
 inputs draw no I

Since no current goes into the - pin those charges flowing thru R must go to the capacitor and with

$$Q = CV_c$$

$$I = \frac{dQ}{dt} = C \frac{dV_c}{dt}$$

The voltage across the capacitor $= V_c = -V_{out}$

$$\text{So } I = -C \frac{dV_{out}}{dt}$$

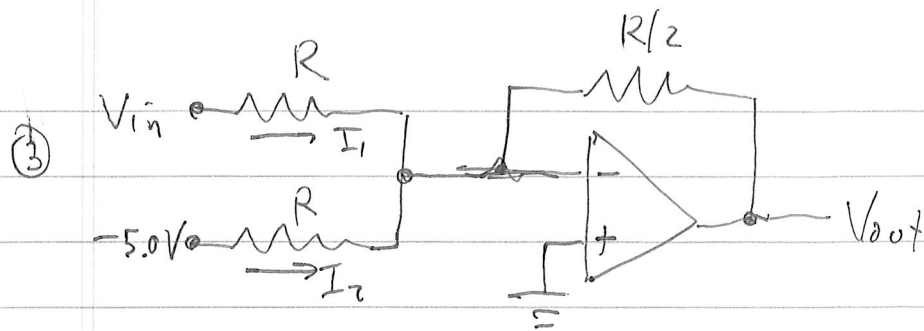
$$\frac{V_{in}}{R} = -C \frac{dV_{out}}{dt} \Rightarrow \frac{dV_{out}}{dt} = -\frac{1}{RC} V_{in}(t)$$

Integrating $\int_{t=0}^t \frac{dV_{out}}{dt'} dt' = -\frac{1}{RC} \int_0^t V_{in}(t') dt'$

$$V_{out}(t) - V_{out}(0) = -\frac{1}{RC} \int_0^t V_{in}(t') dt'$$

Assume $V_{out}(0) = 0$ initially

$$V_{out}(t) = -\frac{1}{RC} \int_0^t V_{in}(t') dt'$$



G.R.: $V_- = V_+ \Rightarrow V_- = 0$ The - pin acts like 'ground'

Then the currents through the input resistors are

$$I_1 = \frac{V_{in}}{R} \quad I_2 = \frac{-5.0V}{R}$$

And these two I 's combine or sum at the - input
G.R. Both must flow around thru $R/2$, not into the V_- pin

So $V_{out} = -\left(\frac{I_1 + I_2}{2}\right) \frac{R}{2}$ (= voltage drop across $\frac{R}{2}$)

$$= -\left(\frac{V_{in}}{R} + \frac{-5.00V}{R}\right) \cdot \frac{R}{2}$$

$$V_{out} = -\left(\frac{5.00V}{2} - \frac{V_{in}}{2}\right) = \frac{V_{in}}{2} - 2.50V$$

If $V_{in} = 0$ $V_{out} = -2.50V$

$V_{in} = -5V$ $V_{out} = \frac{-10V}{2} = -5.0V$

$V_{in} = +5V$ $V_{out} = \frac{5-5V}{2} = 0V$

